

APPLICATION OF THE TLM METHOD TO THE SOUND PROPAGATION MODELLING IN URBAN AREA

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1 ISSUE AND OBJECTIVES

2 TLM METHOD

- TLM method principle
- Homogeneous and non-dissipative atmosphere modelling
- Heterogeneous and dissipative atmosphere modelling
- Boundary condition: pressure reflection coefficient
- Analogy with the wave equation
- Numerical verifications and conclusions

3 IMPEDANCE BOUNDARY CONDITION

- Classical formulation
- Impedance representation
- TLM impedance boundary condition formulation
- Numerical validation

4 VIRTUAL BOUNDARY CONDITION

- TLM absorbing conditions review
- Proposed TLM absorbing layers formulation
- Numerical validation

5 CONCLUSIONS AND OUTLOOK

- Conclusion
- Urban application examples
- Outlook

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Global issue: noise annoyances prevention and abatement

- **health and societal impact of noise**
- **legislative and regulation framework**

LCPC research topic: predicting the noise level in urban environment

- **sound propagation modelling in urban area**

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The urban noise: a complex issue

- street \equiv « opened » waveguide
 - \Rightarrow steady-state phenomena, acoustic « leaks » by the open-tops
- frontages morphology
 - \Rightarrow diffuse reflections, edge diffraction, absorption
- long distance propagation
 - \Rightarrow atmospheric effects, ground effects, « unusual » micrometeorological conditions
- temporal variations
 - \Rightarrow moving/time varying noise sources^[1], micrometeorological conditions fluctuations^[2]

Thesis objective: sound propagation modelling in urban area

- development of a **specific time-domain** numerical model
 - \Rightarrow TLM method (*Transmission Line Modelling*)



[1] A. Can. *Représentation du trafic et caractérisation dynamique du bruit en milieu urbain. PhD Thesis*, Lyon, 2008.



[2] F. Junker *et al.*. *Meteorological classification for environmental acoustics - Practical implications due to experimental accuracy and uncertainty*. ICA, Madrid (Espagne), 2007.

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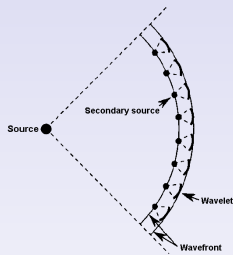
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HUYGENS principle (1690)

A **wavefront** can be broken down into a **set of secondary sources** that radiate spherical wavelets of identical frequency, amplitude and phase.



Numerical adaptation in electromagnetism^[1]

- The **secondary sources** are assimilated to **nodes**.
- The « **diffusion** » of the field between nodes is performed by means of **transmission lines** in term of pulses.



[1] P.B. Johns and R.L. Beurle. *Numerical solution of two dimensional scattering problems using a transmission line matrix. Proc. IEE, 118(9), 1971.*

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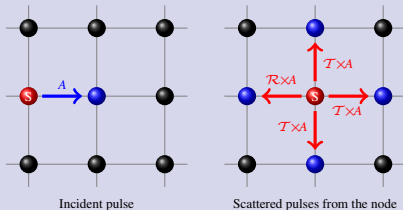
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Simple case in 2D



- Nodal reflection and transmission coefficients:

$$\mathcal{R} = \frac{Z_T - Z_L}{Z_T + Z_L}, \quad \mathcal{R} < 0$$

$$\mathcal{T} = 1 + \mathcal{R} = \frac{2Z_T}{Z_T + Z_L}$$

Z_T : impedance of the termination

Z_L : impedance of the incident transmission line

(here, $Z_L=Z$ and $Z_T=Z/3$, so $\mathcal{R}=-\frac{1}{2}$ and $\mathcal{T}=\frac{1}{2}$)

General case in 2D



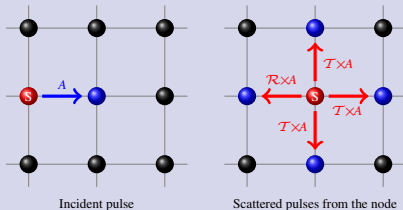
- Matrix relation: ${}_iS = \mathbf{D} \times {}_iI$

where ${}_iI = [{}_iI^1, {}_iI^2, {}_iI^3, {}_iI^4]^T$,

${}_iS = [{}_iS^1, {}_iS^2, {}_iS^3, {}_iS^4]^T$,

and $\mathbf{D} = \begin{bmatrix} \mathcal{R} & \mathcal{T} & \mathcal{T} & \mathcal{T} \\ \mathcal{T} & \mathcal{R} & \mathcal{T} & \mathcal{T} \\ \mathcal{T} & \mathcal{T} & \mathcal{R} & \mathcal{T} \\ \mathcal{T} & \mathcal{T} & \mathcal{T} & \mathcal{R} \end{bmatrix}$.

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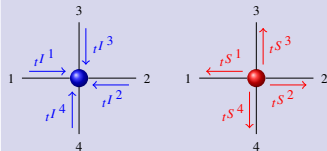
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Z_T : impedance of the termination

Z_L : impedance of the incident transmission line

(here, $Z_L = Z$ and $Z_T = Z/3$, so $\mathcal{R} = -\frac{1}{2}$ and $\mathcal{T} = \frac{1}{2}$)

General case in 2D



- Matrix relation: ${}_t\mathbf{S} = \mathbf{D} \times {}_t\mathbf{I}$

where ${}_t\mathbf{I} = [{}_tI^1, {}_tI^2, {}_tI^3, {}_tI^4]^T$,

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Diffusion in the transmission lines network \Rightarrow connexion laws

$${}_{t+\Delta t}I_{(i,j)}^1 = {}_tS_{(i-1,j)}^2$$

$${}_{t+\Delta t}I_{(i,j)}^2 = {}_tS_{(i+1,j)}^1$$

$${}_{t+\Delta t}I_{(i,j)}^3 = {}_tS_{(i,j-1)}^4$$

$${}_{t+\Delta t}I_{(i,j)}^4 = {}_tS_{(i,j+1)}^3$$

Nodal pressure definition

$${}_tP_{(i,j)} = \frac{1}{2} \sum_{n=1}^4 {}_tI_{(i,j)}^n$$

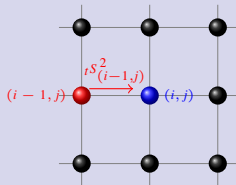
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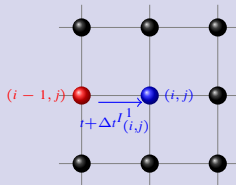
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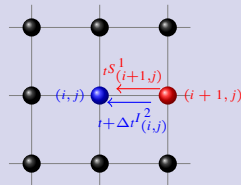
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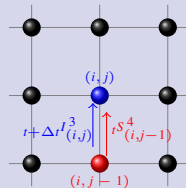
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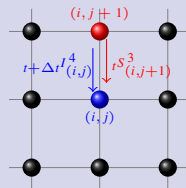
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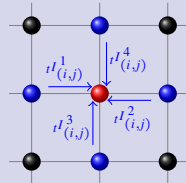
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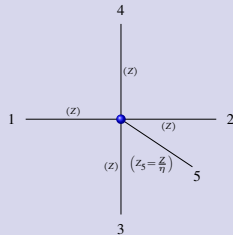
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Heterogeneous propagation medium modelling (micrometeorological conditions)

addition of an open-circuited branch, of impedance Z/η ,
 to the nodal original configuration to introduce
refraction and **turbulence** where the parameter η is
 calculated by^[1]:

$${}_t\eta_{(i,j)} = 4 \left[\left(\frac{c_0}{{}_t c^{\text{eff}}_{(i,j)}} \right)^2 - 1 \right],$$

where ${}_t c^{\text{eff}}_{(i,j)} = \sqrt{\gamma R {}_t T_{(i,j)}} + {}_t \mathbf{W}_{(i,j)} \cdot {}_t \mathbf{u}_{(i,j)}$.



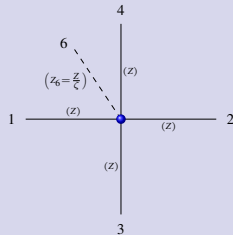
[1] G. Dutilleul. *Applicability of TLM to wind turbine noise prediction*, 2nd Int. Meeting on Wind Turbine Noise, Lyon (France), 2007.

Dissipative propagation medium modelling (atmospheric attenuation)

addition of an anechoic terminated branch, of impedance Z/ζ , to the original nodal configuration where the attenuation factor ζ is defined by^[1]:

$${}_i\zeta_{(i,j)} = -\alpha \sqrt{{}_i\eta_{(i,j)} + 4} \Delta l \frac{\ln(10)}{20},$$

with $\alpha = f(T, P_0, H)$ the **atmospheric absorption coefficient** (expressed in dB.m^{-1}) and Δl the spatial step (in m).



[1] J. Hofmann and K. Heutschi. Numerical simulation of sound wave propagation with sound absorption in time domain. *Appl. Acoust.*, 68(2), 2007.

Heterogeneous and dissipative propagation medium modelling

- Matrix relation: ${}_t\mathbf{S}_{(i,j)} = {}_t\mathbf{D}_{(i,j)} \times {}_t\mathbf{I}_{(i,j)}$,

where ${}_t\mathbf{I}_{(i,j)} = [{}_tI^1; {}_tI^2; {}_tI^3; {}_tI^4; {}_tI^5]^T$,

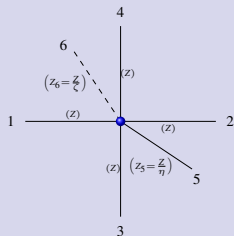
$${}_t\mathbf{S}_{(i,j)} = [{}_tS^1; {}_tS^2; {}_tS^3; {}_tS^4; {}_tS^5]^T,$$

$$\text{and } {}_t\mathbf{D}_{(i,j)} = \frac{2}{{}_t\eta_{(i,j)} + {}_t\zeta_{(i,j)} + 4} \begin{bmatrix} a & 1 & 1 & 1 & \eta \\ 1 & a & 1 & 1 & \eta \\ 1 & 1 & a & 1 & \eta \\ 1 & 1 & 1 & a & \eta \\ 1 & 1 & 1 & 1 & b \end{bmatrix}_{(i,j)}$$

$$\text{with } {}_t a_{(i,j)} = - \left(\frac{{}_t\eta_{(i,j)}}{2} + \frac{{}_t\zeta_{(i,j)}}{2} + 1 \right) \quad \text{and} \quad {}_t b_{(i,j)} = \frac{{}_t\eta_{(i,j)}}{2} - \left(\frac{{}_t\zeta_{(i,j)}}{2} + 2 \right).$$

- Connexion laws: ${}_{t+\Delta t}I^5_{(i,j)} = {}_tS^5_{(i,j)}$

- Nodal pressure: ${}_tP_{(i,j)} = \frac{2}{{}_t\eta_{(i,j)} + {}_t\zeta_{(i,j)} + 4} \left(\sum_{n=1}^4 {}_tI^n_{(i,j)} + {}_t\eta_{(i,j)} {}_tI^5_{(i,j)} \right)$



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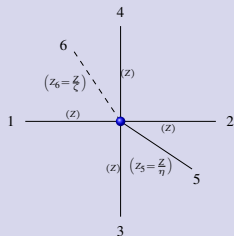
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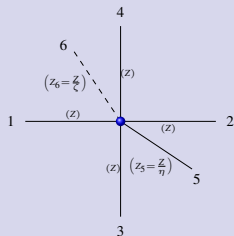
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Wall characterized by a pressure reflection coefficient

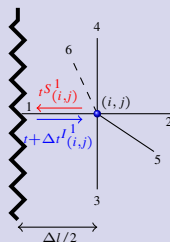
- Example: node (i, j) located at the vicinity of a west side wall defined by a pressure reflection coefficient R_1

$$t+\Delta t I_{(i,j)}^1 = R_1 \times tS_{(i,j)}^1$$

$$t+\Delta t I_{(i,j)}^2 = tS_{(i+1,j)}^1$$

$$t+\Delta t I_{(i,j)}^3 = tS_{(i,j-1)}^4$$

$$t+\Delta t I_{(i,j)}^4 = tS_{(i,j+1)}^3$$



- Relation between the pressure reflection coefficient R_1 and the absorption coefficient in energy α_1 :

$$\alpha_1 = 1 - |R_1|^2$$

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Analogy with the wave equation

- Combination of the matrix relation, the connexion laws and the nodal pressure definition:

$$\frac{{}_i\eta_{(i,j)} + 4}{2} \frac{\Delta t^2}{\Delta l^2} \overbrace{\frac{{}_{i+\Delta t}P(i,j) - 2{}_iP(i,j) + {}_{i-\Delta t}P(i,j)}{\Delta t^2}}^{\partial_{tt}^2 P(i,j)} + {}_i\zeta_{(i,j)} \frac{\Delta t}{\Delta l^2} \overbrace{\frac{{}_{i+\Delta t}P(i,j) - {}_{i-\Delta t}P(i,j)}{2\Delta t}}^{\partial_t P(i,j)} =$$

$$\underbrace{\frac{{}_iP(i+1,j) - 2{}_iP(i,j) + {}_iP(i-1,j)}{\Delta l^2}}_{\partial_{xx}^2 P(i,j)} + \underbrace{\frac{{}_iP(i,j+1) - 2{}_iP(i,j) + {}_iP(i,j-1)}{\Delta l^2}}_{\partial_{yy}^2 P(i,j)}$$

- Helmholtz equation in a heterogeneous and dissipative medium:

$$\left[\Delta + \left(\frac{\omega^2}{c_{\text{TLM}}^2} - j \frac{\omega \zeta_{(i,j)}}{c \Delta l} \right) \right] P_{(i,j)} = 0, \quad c = \frac{\Delta l}{\Delta t}$$

- Celerity correction:

$$c_{\text{TLM}} = \sqrt{\frac{2}{{}_i\eta_{(i,j)} + 4}} c \Rightarrow c = \sqrt{\frac{{}_i\eta_{(i,j)} + 4}{2}} c_0$$

Analogy with the wave equation

- **Combination** of the matrix relation, the connexion laws and the nodal pressure definition:

$$\frac{{}_t\eta(i,j) + 4}{2} \frac{\Delta t^2}{\Delta l^2} \overbrace{\frac{{}_{t+\Delta t}P(i,j) - 2{}_tP(i,j) + {}_{t-\Delta t}P(i,j)}{\Delta t^2}}^{\partial_{tt}^2 P(i,j)} + {}_t\zeta(i,j) \frac{\Delta t}{\Delta l^2} \overbrace{\frac{{}_{t+\Delta t}P(i,j) - {}_{t-\Delta t}P(i,j)}{2\Delta t}}^{\partial_t P(i,j)} =$$

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$$\left[\Delta + \left(\frac{\omega^2}{c_{TLM}^2} - j \frac{\omega \zeta(i,j)}{c \Delta l} \right) \right] P(i,j) = 0, \quad c = \frac{\Delta l}{\Delta t}$$

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$$c_{TLM} = \sqrt{\frac{2}{{}_t\eta(i,j) + 4}} c \Rightarrow c = \sqrt{\frac{{}_t\eta(i,j) + 4}{2}} c_0$$

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Room acoustics applications

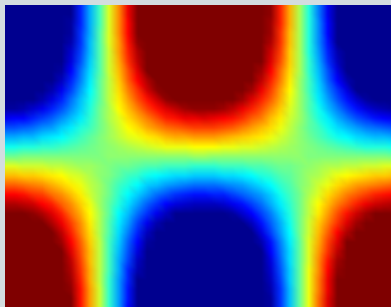


Figure: Eigenmode (2, 1) of a 2D room with perfectly reflecting walls

- dimensions: (6.75 m \times 5.23 m)
- discretization: $\Delta l = 16$ cm and $\Delta t = 0.3$ ms
- sinusoidal source frequency: 60.5 Hz

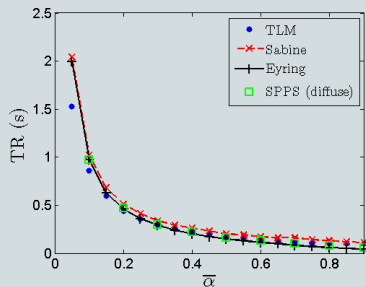
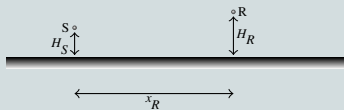


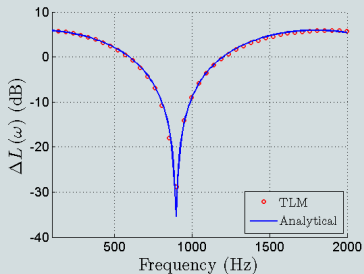
Figure: Reverberation time of a 3D room

- dimensions: (5 m \times 4 m \times 3 m)
- discretization: $\Delta l = 5$ cm and $\Delta t = 8 \times 10^{-5}$ s
- gaussian pulse source frequency: 500 Hz

« Open-space » application



- $H_S = 1$ m, $H_R = 2$ m and $x_R = 20$ m
- perfectly reflective ground
- discretization: $\Delta l = 2$ cm and $\Delta t = 4.1 \times 10^{-5}$ s
- gaussian pulse source frequency: 1500 Hz



Literature study

- few developments
- few validations sometimes limited or even arguable

Thesis contributions

- analytical formulation of a TLM model combining most of the propagative phenomena
- achievement of a generic 2D/3D formulation and numerical implementation
- rigorous validation of the model for academic cases

Main limitations of the model

- no relevant virtual boundary condition formulation in TLM for acoustic modelling
- no realistic boundaries conditions

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5 CONCLUSIONS AND OUTLOOK

- Conclusion
- Urban application examples
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Classical impedance boundary condition formulation:

- in the frequency domain:

$$P_{(b)}(\omega) = Z(\omega) \times V_{n(b)}(\omega)$$

- in the time domain:

$$p_{(b)}(t) = z(t) * v_{n(b)}(t) = \int_{-\infty}^{+\infty} z(t') \times v_{n(b)}(t - t') dt'$$

where $z(t) = \mathfrak{F}^{-1}[Z(\omega)]$

Necessary conditions to transpose $Z(\omega)$ in the time domain:^[1]

- causality
- passivity
- reality



[1] S.W. Rienstra. *Impedance models in time domain including the extended Helmholtz resonator model*. 12th AIAA/CEAS Conf., Cambridge, Massachusetts (USA), 2006.

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1 ISSUE AND OBJECTIVES

2 TLM METHOD

- TLM method principle
- Homogeneous and non-dissipative atmosphere modelling
- Heterogeneous and dissipative atmosphere modelling
- Boundary condition: pressure reflection coefficient
- Analogy with the wave equation
- Numerical verifications and conclusions

3 IMPEDANCE BOUNDARY CONDITION

- Classical formulation
- **Impedance representation**
- TLM impedance boundary condition formulation
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4 VIRTUAL BOUNDARY CONDITION

- TLM absorbing conditions review
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5 CONCLUSIONS AND OUTLOOK

- Conclusion
- Urban application examples
- Outlook

Impedance representation by a sum of 1st order linear systems^[1]

- **in the frequency domain** (response of K linear systems):

$$Z(\omega) = \sum_{k=1}^K \frac{A_k}{\lambda_k - j\omega}$$

where λ_k are real poles ($\lambda_k > 0$)

- **in the time domain** (sum of K impulse responses):

$$z(t) = \sum_{k=1}^K A_k e^{-\lambda_k t} H(t)$$

where $H(t)$ is the HEAVISIDE function



[1] Y. Reymen *et al.*, *Time-domain impedance formulation based on recursive convolution*. 12th AIAA/CEAS Conf., Cambridge, Massachusetts (USA), 2006.

ZWIKKER and KOSTEN impedance model application^[1]

- **Model expression:**

$$Z(\omega) = Z_{\infty} \sqrt{\frac{1 + j\omega\tau}{j\omega\tau}} \quad \text{where } \tau = \frac{\rho_0 q^2 \gamma}{R_S \Omega} \quad \text{a time constant}$$

$$\text{and } Z_{\infty} = \frac{\rho_0 c_0 q}{\Omega} \quad \text{the impedance at the limit } \omega\tau \rightarrow \infty$$

- **Transposition in the time domain^[2]:**

$$z(t) = Z_{\infty} \left[\delta(t) + \frac{1}{\tau} f(\bar{t}) \right] \quad \text{where } \bar{t} = t/\tau$$

- **Impulse response approximation $f(\bar{t})$:**

$$f(\bar{t}) = \frac{e^{-\bar{t}/2}}{2} \left[I_1\left(\frac{\bar{t}}{2}\right) + I_0\left(\frac{\bar{t}}{2}\right) \right] H(\bar{t}) = \sum_{k=1}^K A_k e^{-\lambda_k \bar{t}} H(\bar{t})$$



[1] C. Zwikker and C. W. Kosten. *Sound absorbing materials*. Elsevier Ed., New York, 1949.



[2] V. E. Ostashev *et al.* Padé approximation in time-domain boundary conditions of porous surfaces. *JASA*, 122(1), 2007.

1 ISSUE AND OBJECTIVES

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- Numerical verifications and conclusions

3 IMPEDANCE BOUNDARY CONDITION

- Classical formulation
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- **TLM impedance boundary condition formulation**
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4 VIRTUAL BOUNDARY CONDITION

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5 CONCLUSIONS AND OUTLOOK

- Conclusion
- Urban application examples
- Outlook

Time domain impedance boundary condition formulation^[1]

$$p(m\Delta t) = Z' \left[v_n(m\Delta t) + \sum_{k=1}^K A'_k \psi_k(m\Delta t) \right]$$

where the accumulators ψ_k are given by (\Rightarrow **recursive convolution method**)

$$\psi_k(m\Delta t) = v_n(m\Delta t) \frac{(1 - e^{-\lambda_k \Delta t'})}{\lambda_k} + e^{-\lambda_k \Delta t'} \psi_k((m-1)\Delta t)$$

- ZWIKKER and KOSTEN model: $Z' = Z_\infty$, $A'_k = A_k$ and $\Delta t' = \overline{\Delta t}$
- MIKI model^[2]: $Z' = Z_0$, $A'_k = \frac{A_k \mu}{\Gamma(-b_M)}$ and $\Delta t' = \Delta t$.



[1] Y. Reymen *et al.* *Time-domain impedance formulation based on recursive convolution*. 12th AIAA/CEAS Conf., Cambridge, Massachusetts (USA), 2006.



[2] B. Cotté *Propagation acoustique en milieu extérieur complexe: problèmes spécifiques au ferroviaire dans le contexte des trains à grande vitesse*. *PhD Thesis*, LMFA, École Centrale de Lyon (France), 2008.

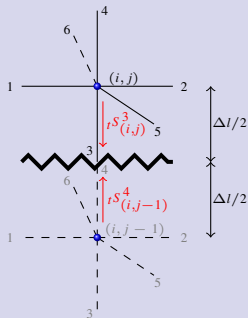
Boundaries modelling in a TLM model: case of the ground

- Introduction of a virtual node
- Boundary **pressure** definition:

$$p\left(t + \frac{\Delta t}{2}\right) = {}_tS^3_{(i,j)} + {}_tS^4_{(i,j-1)}$$

- **Normal particle velocity**:

$$v_n\left(t + \frac{\Delta t}{2}\right) = \frac{{}_tS^3_{(i,j)} - {}_tS^4_{(i,j-1)}}{\rho_0 c}$$



TLM model matching

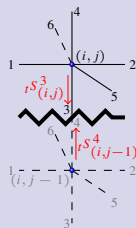
- Scattered pulse from the virtual node:

$$t_m S^4_{(i,j-1)} = t_m S^3_{(i,j)} \left[\frac{-1 + \Lambda_k}{1 + \Lambda_k} \right] + \frac{Z'}{1 + \Lambda_k} \sum_{k=1}^K A'_k e^{-\lambda_k \Delta t'} t_{m-\Delta t} \psi_k$$

$$\text{where } \Lambda_k = \frac{Z'}{\rho_0 c} \left(1 + \sum_{k=1}^K A'_k \frac{1 - e^{-\lambda_k \Delta t'}}{\lambda_k} \right)$$

- Accumulators:

$$t_{m-\Delta t} \psi_k = \left(\frac{t_{m-\Delta t} S^3(i,j) - t_{m-\Delta t} S^4(i,j-1)}{\rho_0 c} \right) \left(\frac{1 - e^{-\lambda_k \Delta t'}}{\lambda_k} \right) + e^{-\lambda_k \Delta t'} t_{m-2\Delta t} \psi_k$$



1 ISSUE AND OBJECTIVES

2 TLM METHOD

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- Homogeneous and non-dissipative atmosphere modelling
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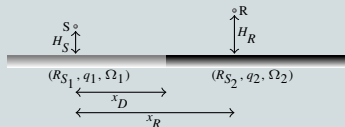
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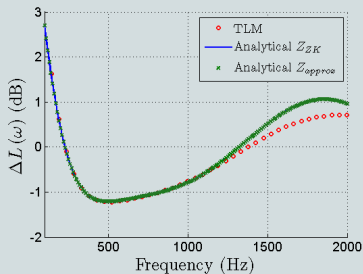
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- Conclusion
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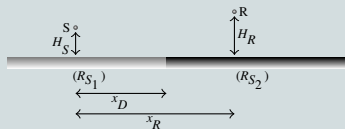
Heterogeneous plane ground: ZWIKKER and KOSTEN impedance model



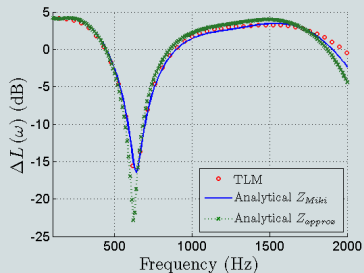
- $H_S = 1$ m, $H_R = 2$ m and $x_R = 20$ m
- discontinuity at $x_D = 10$ m from the source
- $R_{S1} = 10$ kN.s.m⁻⁴, $q_1 = \sqrt{3.5}$ and $\Omega_1 = 0.2$
- $R_{S2} = 100$ kN.s.m⁻⁴, $q_2 = \sqrt{10}$ and $\Omega_2 = 0.5$



Heterogeneous plane ground: MIKI impedance model



- $H_S = 1$ m, $H_R = 2$ m and $x_R = 20$ m
- discontinuity at $x_D = 10$ m from the source
- $R_{S1} = 10$ kN.s.m⁻⁴
- $R_{S2} = 1000$ kN.s.m⁻⁴



1 ISSUE AND OBJECTIVES

2 TLM METHOD

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5 CONCLUSIONS AND OUTLOOK

- Conclusion
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- Outlook

Absorbing boundaries: application of a « non-reflective » termination

- definition of the pressure field on the limit by a **TAYLOR series expansion**^[1]
- application of a **purely real impedance condition**^[2]

Absorbing layers: introduction of an anisotropic absorbing region

• **Perfectly Matched Layers (PML)**



[1] S. El-Masri *et al.*. *Vocal tract acoustics using the transmission line matrix (TLM)*. ICSLP, Philadelphia (USA), 1996.



[2] J. Hofmann and K. Heutschi. *Numerical simulation of sound wave propagation with sound absorption in time domain*. *Appl. Acoust.*, 2007.



[3] D. De Cogan *et al.*. *Transmission Line Matrix in Computational Mechanics*. CRC Press, 2005.

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Absorbing layers: introduction of an anisotropic absorbing region

- Perfectly Matched Layers (PML)^[3]
- modification of the whole connexion laws for the modes located inside the layer^[3]



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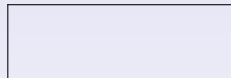
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Proposed absorbing layers formulation

- modification of the connexion law only for the incident pulse propagating in the direction of the computational domain limit

DE COGAN *et al.* formulation

$${}_{t+\Delta t}I_{(i,j)}^1 = \mathbf{F}_{(i,j)} \times {}_tS_{(i-1,j)}^2$$

$${}_{t+\Delta t}I_{(i,j)}^2 = \mathbf{F}_{(i,j)} \times {}_tS_{(i+1,j)}^1$$

$${}_{t+\Delta t}I_{(i,j)}^3 = \mathbf{F}_{(i,j)} \times {}_tS_{(i,j-1)}^4$$

$${}_{t+\Delta t}I_{(i,j)}^4 = \mathbf{F}_{(i,j)} \times {}_tS_{(i,j+1)}^3$$



Proposed formulation

$${}_{t+\Delta t}I_{(i,j)}^1 = \mathbf{F}_{(i,j)} \times {}_tS_{(i-1,j)}^2$$

$${}_{t+\Delta t}I_{(i,j)}^2 = {}_tS_{(i+1,j)}^1$$

$${}_{t+\Delta t}I_{(i,j)}^3 = {}_tS_{(i,j-1)}^4$$

$${}_{t+\Delta t}I_{(i,j)}^4 = {}_tS_{(i,j+1)}^3$$

Proposed absorbing layers formulation

- modification of the connexion law only for the incident pulse propagating in the direction of the computational domain limit

DE COGAN *et al.* formulation

$${}_{t+\Delta t}I_{(i,j)}^1 = \mathbf{F}_{(i,j)} \times {}_tS_{(i-1,j)}^2$$

$${}_{t+\Delta t}I_{(i,j)}^2 = \mathbf{F}_{(i,j)} \times {}_tS_{(i+1,j)}^1$$

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$${}_{t+\Delta t}I_{(i,j)}^4 = \mathbf{F}_{(i,j)} \times {}_tS_{(i,j+1)}^3$$



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Attenuation factor for an absorbing layer of thickness e_{AL}

Looking for a function such as: $\begin{cases} F(d_{(iN,jN)} = 0) = 1 & \text{at the interface} \\ F(d_{(i1,j1)} = e_{AL}) = \epsilon & \text{on the limit, } \epsilon \in]0, 1] \end{cases}$

$$F(d_{(i,j)}) = (1 + \epsilon) - \exp\left[\frac{-(d_{(i,j)} - e_{AL})^2}{B}\right]$$

with $e_{AL} = \frac{\lambda N \lambda_{AL}}{\Delta l}$ and $B = -\frac{e_{AL}^2}{\ln \epsilon}$



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Comparison of the virtual boundary conditions efficiency

$$\text{error}(x, y) = 10 \log_{10} \frac{\sum_{t=0}^T |p_{\text{ff}}(x, y, t) - p(x, y, t)|^2}{\sum_{t=0}^T |p_{\text{ff}}(x, y, t)|^2}$$



Figure: Computational domain

Comparison of the virtual boundary conditions efficiency

$$\text{error}(x, y) = 10 \log_{10} \frac{\sum_{t=0}^T |\mathbf{p}_{\text{ff}}(\mathbf{x}, \mathbf{y}, \mathbf{t}) - p(x, y, t)|^2}{\sum_{t=0}^T |\mathbf{p}_{\text{ff}}(\mathbf{x}, \mathbf{y}, \mathbf{t})|^2}$$

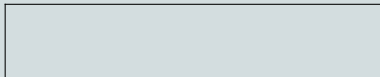


Figure: Free-field computation

Comparison of the virtual boundary conditions efficiency

$$\text{error}(x, y) = 10 \log_{10} \frac{\sum_{t=0}^T |p_{\text{ff}}(x, y, t) - \mathbf{p}(\mathbf{x}, \mathbf{y}, \mathbf{t})|^2}{\sum_{t=0}^T |p_{\text{ff}}(x, y, t)|^2}$$

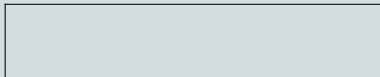


Figure: Virtual free-field computation

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Figure: Virtual free-field computation

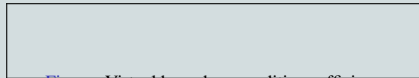


Figure: Virtual boundary conditions efficiency
(AL: $N_{\lambda_{\text{AL}}} = 5$ and $\epsilon = 10^{-5}$)

Urban application

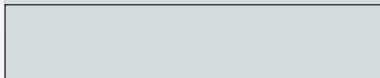


Figure: Street section

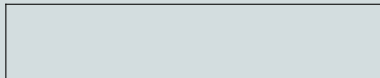


Figure: Sound levels along the receivers axis

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Work done

- analytical formulation and numerical implementation of a 2D/3D TLM model integrating most of the propagative phenomena
- improvement of the method
 - matched impedance boundary condition formulation
 - new formulation of absorbing layers
- validation of the model by comparison with analytical and numerical solutions in academic cases (room acoustics, outdoor sound propagation)

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Parallel streets geometry (quiet street)

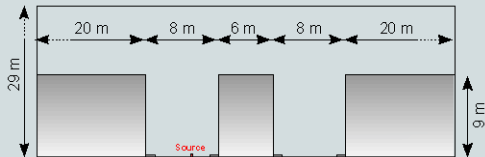


Figure: Gaussian pulse propagation

Urban noise barriers

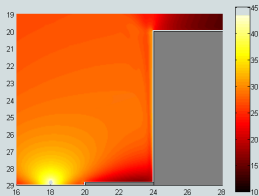


Figure: Without barrier

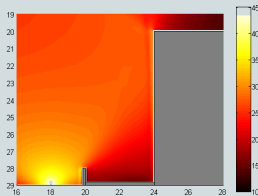


Figure: Green flat barrier

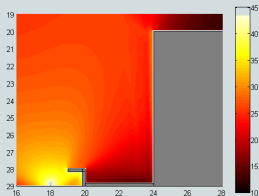


Figure: Perfectly reflective L-shaped barrier

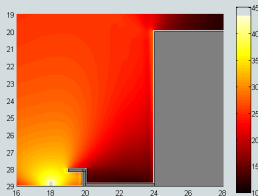


Figure: Green L-shaped barrier

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Outlook concerning our contribution

- thickness consideration in the impedance boundary condition
⇒ coefficients identification in the frequency domain
- rigorous PML formulation for TLM in acoustics

Outlook concerning the TLM model

- atmospheric attenuation frequency dependency
⇒ digital filters^[1]
- sound transmission
⇒ transmission coefficient
⇒ wall acoustic propagation modelling
- tetrahedral 3D mesh^[2]
⇒ 3D simulations with 2D cartesian simulations computational burden
- numerical scheme analysis



[1] T. Tsuchiya. *Numerical simulation of sound wave propagation with sound absorption in time domain*. 13th Int. Cong. Sound Vib., Vienne, 2006.



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Outlook in terms of validation

- micrometeorological conditions implementation
- comparison with experimental results

Outlook in terms of applications

- auralization (soundscape virtual modelling)
- coupling with road traffic models
- ...

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Thank you for your attention

This thesis' work is supported by the following organizations' scientific drafts:

- LCPC - Opération 11M061: « *Prévoir le bruit en milieu urbain* » (« *Forecast the noise level in urban environment* »)
- IRSTV CNRS 2488, PRF « *Environnements sonores urbains* » (« *Sound urban environments* »)
- GdR CNRS 2493, thème 2: « *Propagation en espace urbain et en milieu ouvert* » (« *Propagation in urban area and in open-space* »)

Wave propagation equations in absorbing layers

- theoretical wave propagation equation in PML^[1]:

$$\frac{1}{c_0^2} \frac{\partial^2 p}{\partial t^2} - \frac{\partial^2 p}{\partial x^2} = -\frac{1}{c_0^2} q_x \frac{\partial p}{\partial t} + \rho_0 q_x \frac{\partial u_x}{\partial x} + \rho_0 u_x \frac{\partial q_x}{\partial x}$$

- discrete wave propagation equation obtained with the proposed method:

$$\frac{\Delta t^2}{\Delta l^2} \frac{{}_t P_{(i)} - 2{}_t P_{(i)} + {}_{t-\Delta t} P_{(i)}}{\Delta t^2} - \frac{{}_t P_{(i+1)} - 2{}_t P_{(i)} + {}_t P_{(i-1)}}{\Delta l^2} =$$

$$-F_{(i)} \frac{\Delta t^2}{\Delta l^2} \frac{{}_t P_{(i)} - {}_{t-\Delta t} P_{(i)}}{\Delta t} + \rho_0 F_{(i)} \frac{{}_t u_{(i+1)} - {}_t u_{(i)}}{\Delta l} + \rho_0 {}_t u_{(i)} \frac{F_{(i+1)} - F_{(i)}}{\Delta l} + \frac{\Theta}{\Delta l^2}$$

$$\text{with } \Theta = -F_{(i+1)} \left[{}_t S_{(i)}^2 - {}_t S_{(i)}^1 \right] + [F_{(i)} - 1] {}_{t-\Delta t} S_{(i)}^1 - 2F_{(i)} {}_t S_{(i-1)}^2$$

$$- [F_{(i+1)} - 1] {}_{t-\Delta t} S_{(i)}^2 + {}_t S_{(i)}^1 + {}_t S_{(i)}^2 - {}_{t-\Delta t} S_{(i)}^1.$$



[1] Q. Qi and T.L. Geers. *Evaluation of the perfectly matched layer for computational acoustics*. *J. Comput. Phys.*, 139(1), 1997.